

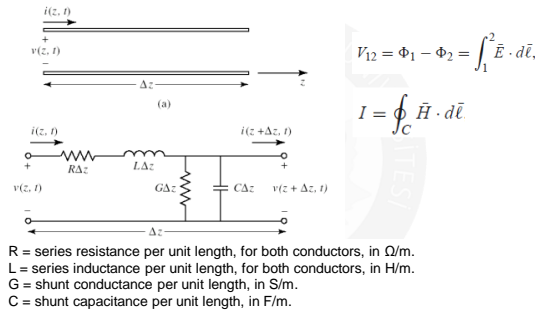
Microwave Techniques

Transmission Lines

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Autumn 2016

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Lumped Element Model for a Transmission Line

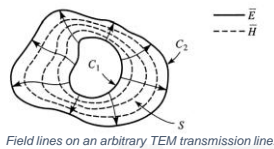


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Field Analysis of Transmission Lines

Transmission Line Parameters



$$R = \frac{R_s}{|I_o|^2} \int_{C_1+C_2} \vec{H} \cdot \vec{H}^* dl \quad \Omega/\text{m}.$$

$$G = \frac{\omega \epsilon''}{|V_o|^2} \int_S \vec{E} \cdot \vec{E}^* ds \quad \text{S}/\text{m}.$$

$$W_m = \frac{\mu}{4} \int_S \vec{H} \cdot \vec{H}^* ds = L |I_o|^2 / 4$$

$$L = \frac{\mu}{|I_o|^2} \int_S \vec{H} \cdot \vec{H}^* ds \quad \text{H}/\text{m}$$

$$W_e = \frac{\epsilon}{4} \int_S \vec{E} \cdot \vec{E}^* ds = C |V_o|^2 / 4$$

$$C = \frac{\epsilon}{|V_o|^2} \int_S \vec{E} \cdot \vec{E}^* ds \quad \text{F}/\text{m}.$$

$R_s = 1/\sigma \delta_s$
is the surface resistance of the conductors
 ϵ''
is the imaginary part of the complex permittivity

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Transmission Line Parameters of a Coaxial Line

Geometry of a coaxial line with surface resistance R_s on the inner and outer conductors.

$$\vec{E} = \frac{V_o \hat{\rho}}{\rho \ln b/a} e^{-\gamma z},$$

$$\vec{H} = \frac{I_o \hat{\phi}}{\gamma \pi a} e^{-\gamma z},$$

$$L = \frac{\mu}{(2\pi)^2} \int_{\phi=0}^{2\pi} \int_{\rho=a}^b \frac{1}{\rho^2} \rho d\rho d\phi = \frac{\mu}{2\pi} \ln b/a \quad \text{H}/\text{m}$$

$$C = \frac{\epsilon'}{(\ln b/a)^2} \int_{\phi=0}^{2\pi} \int_{\rho=a}^b \frac{1}{\rho^2} \rho d\rho d\phi = \frac{2\pi \epsilon'}{\ln b/a} \quad \text{F}/\text{m}$$

$$R = \frac{R_s}{(2\pi)^2} \left\{ \int_{\phi=0}^{2\pi} \frac{1}{a^2} a d\phi + \int_{\phi=0}^{2\pi} \frac{1}{b^2} b d\phi \right\} = \frac{R_s}{2\pi} \left(\frac{1}{a} + \frac{1}{b} \right) \quad \Omega/\text{m}$$

$$G = \frac{\omega \epsilon''}{(\ln b/a)^2} \int_{\phi=0}^{2\pi} \int_{\rho=a}^b \frac{1}{\rho^2} \rho d\rho d\phi = \frac{2\pi \omega \epsilon''}{\ln b/a} \quad \text{S}/\text{m}$$

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Transmission Line Parameters of Some Common Lines

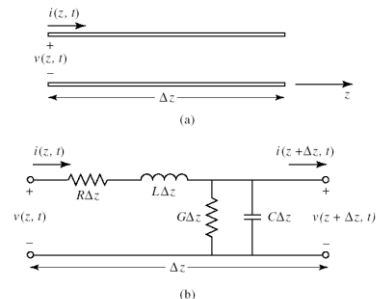
TABLE 2.1 Transmission Line Parameters for Some Common Lines

	COAX	TWO-WIRE	PARALLEL-PLATE
L	$\frac{\mu}{2\pi} \ln \frac{b}{a}$	$\frac{\mu}{\pi} \cosh^{-1} \left(\frac{D}{2a} \right)$	$\frac{\mu d}{w}$
C	$\frac{2\pi \epsilon'}{\ln b/a}$	$\frac{\pi \epsilon'}{\cosh^{-1}(D/2a)}$	$\frac{\epsilon' w}{d}$
R	$\frac{R_s}{2\pi} \left(\frac{1}{a} + \frac{1}{b} \right)$	$\frac{R_s}{\pi a}$	$\frac{2R_s}{w}$
G	$\frac{2\pi \omega \epsilon''}{\ln b/a}$	$\frac{\pi \omega \epsilon''}{\cosh^{-1}(D/2a)}$	$\frac{\omega \epsilon'' w}{d}$

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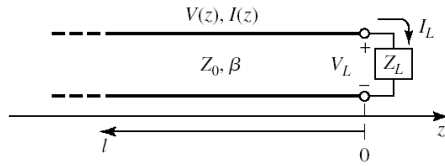
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Terminated Lossless Transmission Lines

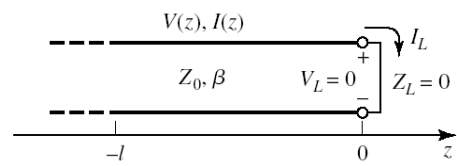


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Short-Circuit Termination

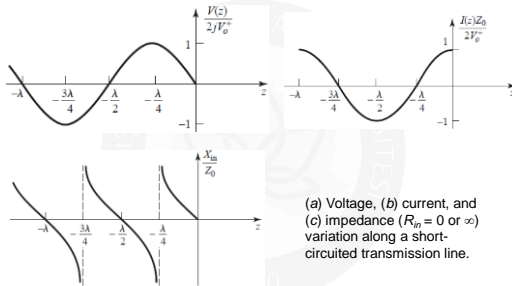


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Short-Circuit Termination

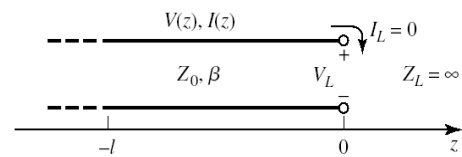


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Open-Circuit Termination

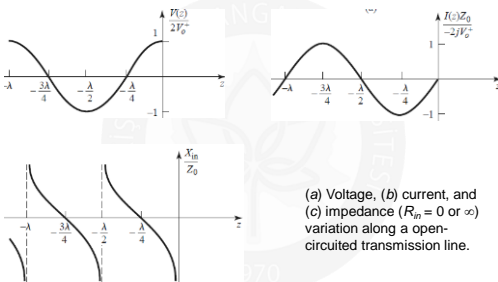


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Open-Circuit Termination

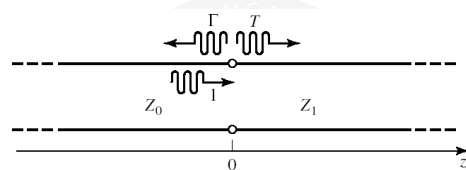


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Junction of Two Transmission Lines



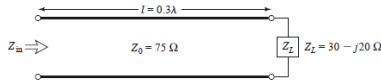
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Examples

- 2.1 A $75\ \Omega$ coaxial line has a current $i(t, z) = 1.8 \cos(3.77 \times 10^9 t - 18.13z)$ mA. Determine (a) the frequency, (b) the phase velocity, (c) the wavelength, (d) the relative permittivity of the line, (e) the phasor form of the current, and (f) the time domain voltage on the line.
- 2.2 A transmission line has the following per-unit-length parameters: $L = 0.5\ \mu\text{H/m}$, $C = 200\ \text{pF/m}$, $R = 4.0\ \Omega/\text{m}$, and $G = 0.02\ \text{S/m}$. Calculate the propagation constant and characteristic impedance of this line at $800\ \text{MHz}$. If the line is $30\ \text{cm}$ long, what is the attenuation in dB? Recalculate these quantities in the absence of loss ($R = G = 0$).
- 2.8 A lossless transmission line of electrical length $\ell = 0.3\lambda$ is terminated with a complex load impedance as shown in the accompanying figure. Find the reflection coefficient at the load, the SWR on the line, the reflection coefficient at the input of the line, and the input impedance to the line.



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Smith Chart

$$\Gamma = \frac{z_L - 1}{z_L + 1} = |\Gamma|e^{j\theta} \quad z_L = \frac{1 + |\Gamma|e^{j\theta}}{1 - |\Gamma|e^{j\theta}}$$

$$r_L = \frac{1 - \Gamma_r^2 - \Gamma_i^2}{(1 - \Gamma_r)^2 + \Gamma_i^2}, \quad x_L = \frac{2\Gamma_i}{(1 - \Gamma_r)^2 + \Gamma_i^2}$$

$$r_L + jx_L = \frac{(1 + \Gamma_r) + j\Gamma_i}{(1 - \Gamma_r) - j\Gamma_i}$$

$$\left(\Gamma_r - \frac{r_L}{1 + r_L}\right)^2 + \Gamma_i^2 = \left(\frac{1}{1 + r_L}\right)^2$$

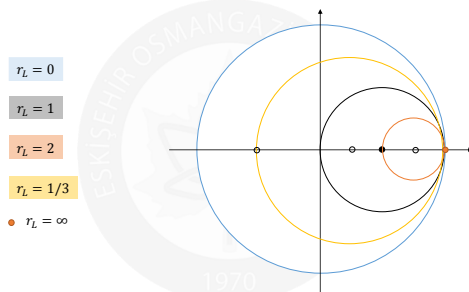
$$(\Gamma_r - 1)^2 + \left(\Gamma_i - \frac{1}{x_L}\right)^2 = \left(\frac{1}{x_L}\right)^2$$

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Smith Chart

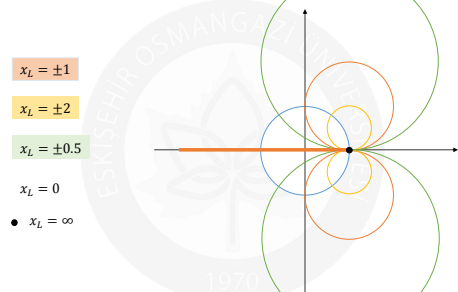


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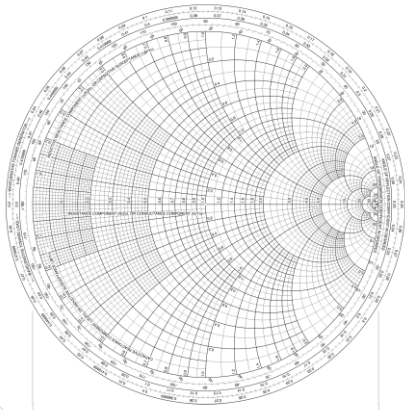


Smith Chart



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Examples

EXAMPLE 2.2

A load impedance of $40 + j70\ \Omega$ terminates a $100\ \Omega$ transmission line that is 0.3λ long. Find the reflection coefficient at the load, the reflection coefficient at the input to the line, the input impedance, the standing wave ratio on the line, and the return loss.

EXAMPLE 2.3

A load of $Z_L = 100 + j50\ \Omega$ terminates a $50\ \Omega$ line. What are the load admittance and input admittance if the line is 0.15λ long?

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