

Introduction to Laplace Transforms

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Introduction

- Laplace transform:
 - important method of representing circuits and signals
 - used to analyze circuits with complicated signals
- To analyze circuits using Laplace transform:
 - change signals from time-domain to frequency domain
 - analyze circuits in frequency domain
 - convert signals in frequency-domain back to time domain
- More general than phasor method



Definition of the Laplace Transform

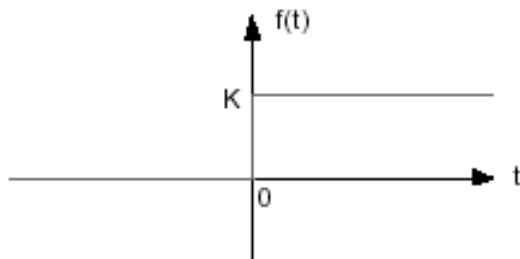
One-sided Laplace transform of a time function $f(t)$ is defined as

$$F(s) = L\{f(t)\} = \int_{0^-}^{\infty} f(t)e^{-st} dt$$

- Some functions may not have Laplace transforms but we do not use them in circuit analysis.
- $F(s)$ is a one sided transform (starts from $t=0$).
- Choose 0^- as the lower limit (to capture discontinuity in $f(t)$ due to an event such as closing a switch).



The Step Function



if $K = 1$, unit step function is defined as

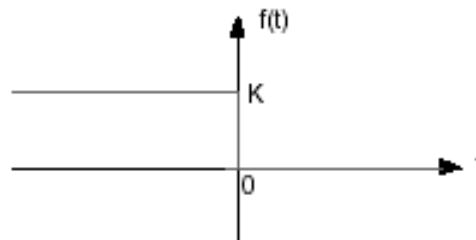
$$u(t) = \begin{cases} 0 & t < 0 \\ 1 & t > 0 \end{cases}$$

if $K \neq 1$, the step function is defined as

$$Ku(t) = \begin{cases} 0 & t < 0 \\ K & t > 0 \end{cases}$$

Note: the step function is not defined for $t=0$.

$$Ku(-t) = \begin{cases} K & t < 0 \\ 0 & t > 0 \end{cases}$$





Laplace Transform of the Step Function

$$\begin{aligned} L\{u(t)\} &= \int_{0^-}^{\infty} u(t)e^{-st} dt = \int_0^{\infty} e^{-st} dt \\ &= -\frac{1}{s} e^{-st} \Big|_0^{\infty} = \frac{1}{s} \end{aligned}$$



The Impulse Function

Impulse = signal of infinite amplitude and zero duration

the unit impulse function :

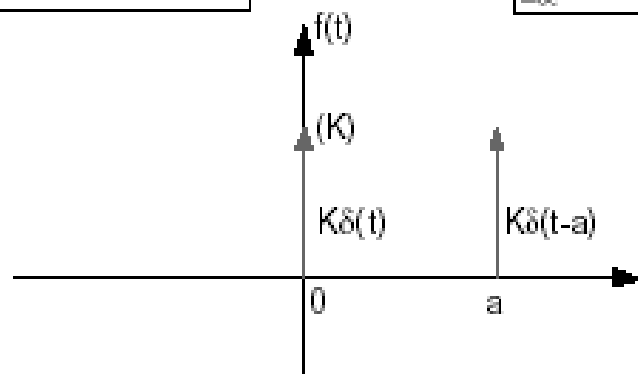
$$\delta(t) = 0 \quad t \neq 0$$

$$\int_{-\infty}^{\infty} \delta(t) dt = 1$$

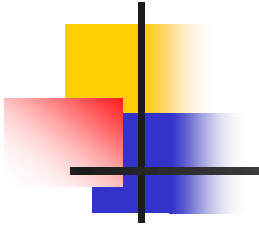
the impulse function :

$$K\delta(t) = 0 \quad t \neq 0$$

$$\int_{-\infty}^{\infty} K\delta(t) dt = K$$



K – strength of
the impulse
function



The sifting property: $\int_{-\infty}^{+\infty} f(t)\delta(t-a)dt = f(a)$

The impulse function is a derivative of the step function:

$$\delta(t) = \frac{du(t)}{dt} \qquad \int_{-\infty}^t \delta(x)dx = u(t)$$

The Laplace transform of the impulse function:

$$L\{\delta(t-a)\} = \int_{0^-}^{\infty} \delta(t-a)e^{-st}dt = e^{-as}$$

$$\text{For } a = 0, \quad L\{\delta(t)\} = 1$$



Operational Transforms

1) Multiplication by a constant

$$\text{If } L\{f(t)\} = F(s) \quad \text{then } L\{Kf(t)\} = KF(s)$$

2) Addition (subtraction)

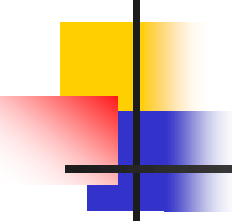
$$\begin{aligned} \text{If } L\{f_1(t)\} = F_1(s), \quad L\{f_2(t)\} = F_2(s), \quad L\{f_3(t)\} = F_3(s) \\ \text{then } L\{f_1(t) + f_2(t) - f_3(t)\} = F_1(s) + F_2(s) - F_3(s) \end{aligned}$$

3) Differentiation

$$L\left\{\frac{df(t)}{dt}\right\} = sF(s) - f(0^-), \quad L\left\{\frac{d^2 f(t)}{dt^2}\right\} = s^2 F(s) - sf(0^-) - f'(0^-)$$

$$L\left\{\frac{d^n f(t)}{dt^n}\right\} = s^n F(s) - s^{n-1} f(0^-) - s^{n-2} f'(0^-) - \dots - sf^{(n-2)}(0^-) - f^{(n-1)}(0^-)$$





4) Integration $L\left\{\int_{0^-}^t f(x)dx\right\} = \frac{F(s)}{s}$

5) Translation in the Time Domain

$$L\{f(t-a)u(t-a)\} = e^{-as}F(s), \quad a \succ 0$$

6) Translation in the Frequency Domain

$$L\{e^{-at}f(t)\} = F(s+a)$$

7) Scale Changing

$$L\{f(at)\} = \frac{1}{a}F\left(\frac{s}{a}\right), \quad a \succ 0$$



Laplace Transform of Cos and Sine

$$\begin{aligned} L\{\cos \omega t u(t)\} &= \int_{0^-}^{\infty} \cos \omega t u(t) e^{-st} dt = \int_0^{\infty} \cos \omega t e^{-st} dt \\ &= \int_0^{\infty} \frac{e^{j\omega t} + e^{-j\omega t}}{2} e^{-st} dt = \frac{1}{2} \left[\int_0^{\infty} e^{-(s-j\omega)t} dt + \int_0^{\infty} e^{-(s+j\omega)t} dt \right] \\ &= -\frac{1}{2} \frac{1}{s-j\omega} e^{-(s-j\omega)t} \bigg|_0^{\infty} - \frac{1}{2} \frac{1}{s+j\omega} e^{-(s+j\omega)t} \bigg|_0^{\infty} \\ &= \frac{1}{2} \frac{1}{s-j\omega} + \frac{1}{2} \frac{1}{s+j\omega} = \frac{s}{s^2 + \omega^2} \end{aligned}$$

$$L\{\sin \omega t u(t)\} = \frac{\omega}{s^2 + \omega^2}$$



Inverse Laplace Transform

- Goal: find the inverse Laplace transform of a function that has the following form:

$$F(s) = \frac{N(s)}{D(s)} = \frac{a_n s^n + a_{n-1} s^{n-1} + \dots + a_1 s + a_0}{b_m s^m + b_{m-1} s^{m-1} + \dots + b_1 s + b_0}$$

- If $m > n$
 - $F(s)$ is a proper rational function

$$\text{example : } F(s) = \frac{(s-4)(s+4)}{s(s-6)(s^2+6s+25)}$$

- If $m \leq n$
 - $F(s)$ is an improper rational function

$$\text{example : } F(s) = \frac{(s-4)(s+4)(s^2+6s+25)}{s(s-6)}$$



Partial Fraction Expansion

Step 1: Expand $F(s)$ as a sum of partial fractions.

Step 2: Compute the expansion constants (four cases)

Step 3: Write the inverse transform

$$\frac{s+6}{s(s+3)(s+1)^2}$$

Denominator has four roots: $s=0$ and $s=-3$ are distinct. $s=-1$ is a multiple root of multiplicity 2.

$$\frac{s+6}{s(s+3)(s+1)^2} \equiv \frac{K_1}{s} + \frac{K_2}{s+3} + \frac{K_3}{(s+1)^2} + \frac{K_4}{s+1}$$

$$\begin{aligned} & L^{-1} \left\{ \frac{s+6}{s(s+3)(s+1)^2} \right\} \\ &= (K_1 + K_2 e^{-3t} + K_3 t e^{-t} + K_4 e^{-t}) u(t) \end{aligned}$$



Distinct Real Roots

$$F(s) = \frac{96(s+5)(s+12)}{s(s+8)(s+6)} = \frac{K_1}{s} + \frac{K_2}{s+8} + \frac{K_3}{s+6}$$

$$K_1 = sF(s)\Big|_{s=0} = s \frac{96(s+5)(s+12)}{s(s+8)(s+6)} \Big|_{s=0} = \frac{96(5)(12)}{8(6)} = 120$$

$$K_2 = (s+8)F(s)\Big|_{s=-8} = (s+8) \frac{96(s+5)(s+12)}{s(s+8)(s+6)} \Big|_{s=-8} = \frac{96(-3)(4)}{-8(-2)} = -72$$

$$K_3 = (s+6)F(s)\Big|_{s=-6} = (s+6) \frac{96(s+5)(s+12)}{s(s+8)(s+6)} \Big|_{s=-6} = \frac{96(-1)(6)}{-6(2)} = 48$$

$$F(s) = \frac{120}{s} + \frac{48}{s+6} - \frac{72}{s+8}$$

$$f(t) = (120 + 48e^{-6t} - 72e^{-8t})u(t)$$



Distinct Complex Roots

$$F(s) = \frac{100(s+3)}{(s+6)(s^2+6s+25)} = \frac{100(s+3)}{(s+6)(s+3+j4)(s+3-j4)}$$

$$= \frac{K_1}{s+6} + \frac{K_2}{s+3-j4} + \frac{K_3}{s+3+j4}$$

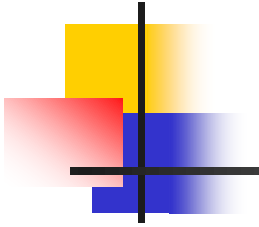
$$K_1 = \left. \frac{100(s+3)}{s^2+6s+25} \right|_{s=-6} = \frac{100(-3)}{25} = -12$$

$$K_2 = \left. \frac{100(s+3)}{(s+6)(s+3+j4)} \right|_{s=-3+j4} = \frac{100(j4)}{(3+j4)(j8)}$$

$$= 6 - j8 = 10e^{-j53.13^\circ}$$

$$K_3 = \left. \frac{100(s+3)}{(s+6)(s+3-j4)} \right|_{s=-3-j4} = \frac{100(-j4)}{(3-j4)(-j8)}$$

$$= 6 + j8 = 10e^{j53.13^\circ}$$



$$F(s) = \frac{-12}{s+6} + \frac{10 \angle -53.13^\circ}{s+3-j4} + \frac{10 \angle 53.13^\circ}{s+3+j4}$$

$$f(t) = (-12e^{-6t} + 10e^{-j53.13^\circ} e^{-(3-j4)t} + 10e^{j53.13^\circ} e^{-(3+j4)t})u(t)$$

$$10e^{-j53.13^\circ} e^{-(3-j4)t} + 10e^{j53.13^\circ} e^{-(3+j4)t}$$

$$= 10e^{-3t} (e^{j(4t-53.13^\circ)} + e^{-j(4t-53.13^\circ)})$$

$$= 20e^{-3t} \cos(4t - 53.13^\circ)$$

$$f(t) = [-12e^{-6t} + 20e^{-3t} \cos(4t - 53.13^\circ)]u(t)$$



Whenever $F(s)$ contains distinct complex roots at the denominator as $(s+\alpha-j\beta)(s+\alpha+j\beta)$, a pair of terms of the form

$$\frac{K}{s+\alpha-j\beta} + \frac{K^*}{s+\alpha+j\beta} \quad \text{appears in the partial fraction.}$$

Where K is a complex number in polar form $K=|K|e^{j\theta}=|K|\angle\theta^\circ$ and K^* is the complex conjugate of K .

The inverse Laplace transform of the complex-conjugate pair is

$$\begin{aligned} L^{-1} &= \left\{ \frac{K}{s+\alpha-j\beta} + \frac{K^*}{s+\alpha+j\beta} \right\} \\ &= 2|K|e^{-\alpha t} \cos(\beta t + \theta) \end{aligned}$$



Repeated Real Roots

$$F(s) = \frac{100(s+25)}{s(s+5)^3} = \frac{K_1}{s} + \frac{K_2}{(s+5)^3} + \frac{K_3}{(s+5)^2} + \frac{K_4}{s+5}$$

$$K_1 = \left. \frac{100(s+25)}{(s+5)^3} \right|_{s=0} = \frac{100(25)}{125} = 20$$

$$K_2 = \left. \frac{100(s+25)}{s} \right|_{s=-5} = \frac{100(20)}{-5} = -400$$

$$K_3 = \frac{d}{ds} [(s+5)^3 F(s)]_{s=-5} = 100 \left[\frac{s-(s+25)}{s^2} \right]_{s=-5} = -100$$

$$K_4 = \frac{1}{2} \frac{d^2}{ds^2} [(s+5)^3 F(s)]_{s=-5} = \frac{1}{2} 100 \left[\frac{-2s(25)}{s^4} \right]_{s=-5} = -20$$

$$F(s) = \frac{20}{s} - \frac{400}{(s+5)^3} - \frac{100}{(s+5)^2} - \frac{20}{s+5}$$

$$f(t) = [20 - 200t^2 e^{-5t} - 100te^{-5t} - 20e^{-5t}]u(t)$$

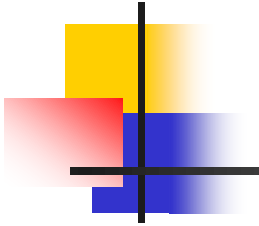


Repeated Complex Roots

$$F(s) = \frac{768}{(s^2 + 6s + 25)^2} = \frac{768}{(s + 3 - j4)^2 (s + 3 + j4)^2}$$
$$= \frac{K_1}{(s + 3 - j4)^2} + \frac{K_1^*}{(s + 3 + j4)^2} + \frac{K_2}{(s + 3 - j4)} + \frac{K_2^*}{(s + 3 + j4)}$$

$$K_1 = \left. \frac{768}{(s + 3 + j4)^2} \right|_{s=-3+j4} = \frac{768}{(j8)^2} = -12$$

$$K_2 = \left. \frac{d}{ds} \left[\frac{768}{(s + 3 + j4)^2} \right] \right|_{s=-3+j4} = - \left. \frac{2(768)}{(s + 3 + j4)^3} \right|_{s=-3+j4}$$
$$= - \frac{2(768)}{(j8)^3} = -j3 = 3 \angle -90^\circ$$



$$F(s) = \left[\frac{-12}{(s+3-j4)^2} + \frac{-12}{(s+3+j4)^2} \right] + \left(\frac{3 \angle -90^\circ}{s+3-j4} + \frac{3 \angle 90^\circ}{s+3+j4} \right)$$

$$f(t) = [-24te^{-3t} \cos 4t + 6e^{-3t} \cos(4t - 90^\circ)]u(t)$$



Improper Transfer Functions

An improper transfer function can always be expanded into a polynomial plus a proper transfer function.

$$F(s) = \frac{s^4 + 13s^3 + 66s^2 + 200s + 300}{s^2 + 9s + 20}$$

$$= s^2 + 4s + 10 + \frac{30s + 100}{s^2 + 9s + 20}$$

$$= s^2 + 4s + 10 + \frac{-20}{s + 4} + \frac{50}{s + 5}$$

$$f(t) = \frac{d^2\delta(t)}{dt^2} + 4\frac{d\delta(t)}{dt} + 10\delta(t) + (-20e^{-4t} + 50e^{-5t})u(t)$$

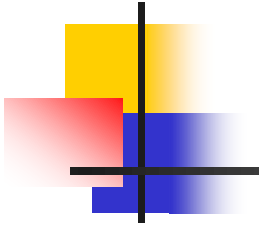


POLES AND ZEROS OF $F(s)$

The rational function $F(s)$ may be expressed as the ratio of two factored polynomials as

$$F(s) = \frac{K(s + z_1)(s + z_2) \cdots (s + z_n)}{(s + p_1)(s + p_2) \cdots (s + p_m)}$$

The roots of the denominator polynomial $-p_1, -p_2, \dots, -p_m$ are called the poles of $F(s)$. At these values of s , $F(s)$ becomes infinitely large. The roots of the numerator polynomial $-z_1, -z_2, \dots, -z_n$ are called the zeros of $F(s)$. At these values of s , $F(s)$ becomes zero.

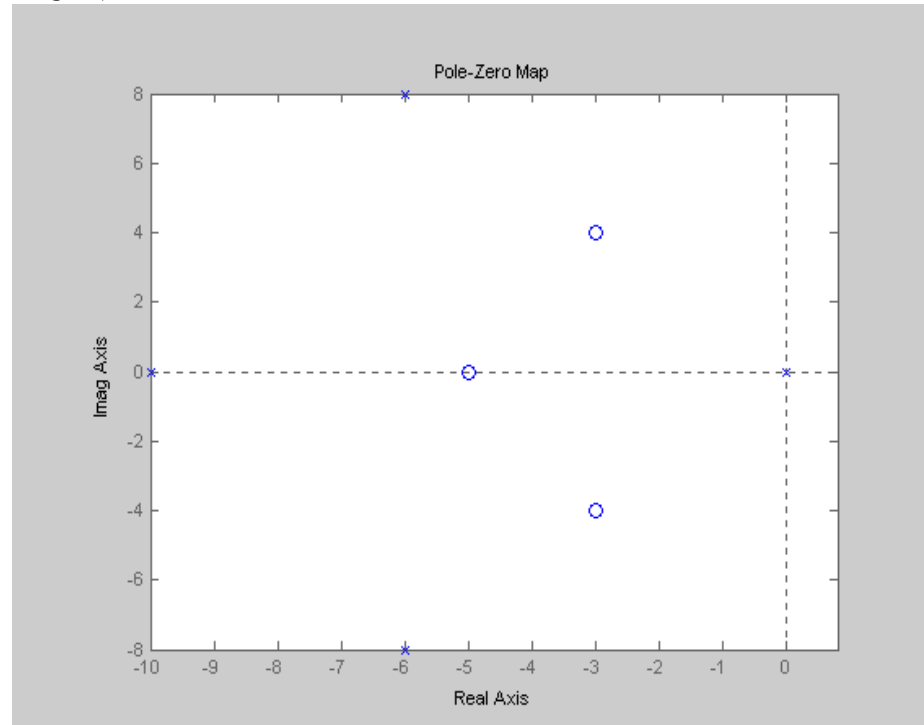


$$F(s) = \frac{10(s+5)(s+3-j4)(s+3+j4)}{s(s+10)(s+6-j8)(s+6+j8)}$$

The poles of $F(s)$ are at
0, -10, -6+j8, and -6-j8.

The zeros of $F(s)$
are at -5, -3+j4, -3-j4

```
num=conv([1 5],[1 6 25]);  
den=conv([1 10 0],[1 12 100]);  
pzmap(num,den)
```





Initial-Value Theorem

The initial-value theorem enables us to determine the value of $f(t)$ at $t=0$ from $F(s)$. This theorem assumes that $f(t)$ contains no impulse functions and poles of $F(s)$, except for a first-order pole at the origin, lie in the left half of the s plane.

$$\lim_{t \rightarrow 0^+} f(t) = \lim_{s \rightarrow \infty} sF(s)$$

$$F(s) = \frac{100(s+3)}{(s+6)(s^2+6s+25)}$$

$$f(t) = [-12e^{-6t} + 20e^{-3t} \cos(4t - 53.13^\circ)]u(t)$$



Final-Value Theorem

The final-value theorem enables us to determine the behavior of $f(t)$ at infinity using $F(s)$.

$$\lim_{t \rightarrow \infty} f(t) = \lim_{s \rightarrow 0} sF(s)$$

The final-value theorem is useful only if $f(\infty)$ exists. This condition is true only if all the poles of $F(s)$, except for a simple pole at the origin, lie in the left half of the s plane.

$$\lim_{s \rightarrow 0} F(s) = \lim_{s \rightarrow 0} s \frac{100(s+3)}{(s+6)(s^2+6s+25)} = 0$$

$$\lim_{t \rightarrow \infty} f(t) = \lim_{t \rightarrow \infty} [-12e^{-6t} + 20e^{-3t} \cos(4t - 53.13^\circ)]u(t) = 0$$